## General Description

The MIC2095/97/98/99 family of switches are selfcontained, current-limiting, high-side power switches, ideal for power-control applications. These switches are useful for general purpose power distribution applications such as digital televisions (DTV), printers, set-top boxes (STB), PCs, PDAs, and other peripheral devices.
The current limiting switches feature either a fixed $0.5 \mathrm{~A} / 0.9 \mathrm{~A}$ or resistor programmable output current limit. The family also has fault blanking to eliminate false noiseinduced, over current conditions. After an over-current condition, these devices automatically restart if the enable pin remains active. The MIC2097 switch offers a unique new patented Kickstart feature, which allows momentary high-current surges up to the secondary current limit (lumit_2nd). This is useful for charging loads with high inrush currents, such as capacitors.
The MIC2095/97/98/99 family of switches provides undervoltage, over-temperature shutdown, and output fault status reporting. The family also provides either an active low or active high, logic level enable pin.
The MIC2095/97/98/99 family is offered in a space saving $1.6 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ Thin MLF ${ }^{\circledR}$ (TMLF) package.
Datasheets and support documentation can be found on Micrel's web site at: www.micrel.com.

## Features

- MIC2095: 0.5A fixed current limit
- MIC2098: 0.9A fixed current limit
- MIC2097/99: Resistor programmable current limit - 0.1A to 1.1A
- MIC2097: Kickstart for high peak current loads
- Under voltage lock-out (UVLO)
- Soft start prevents large current inrush
- Automatic-on output after fault
- Thermal protection
- Enable active high or active low
- $170 \mathrm{~m} \Omega$ typical on-resistance @ 5 V
- $2.5 \mathrm{~V}-5.5 \mathrm{~V}$ operating range


## Applications

- Digital televisions (DTV)
- Set top boxes
- PDAs
- Printers
- USB / IEEE 1394 power distribution
- Desktop and laptop PCs
- Game consoles
- USB keyboard
- Docking stations


## Typical Application



MIC2095 USB Power Switch

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## Ordering Information

| Part Number | Marking | ENABLE <br> Logic | Kickstart ${ }^{(\text {Tm })}$ | ILIMIT | FAULT/ <br> Output | Junction <br> Temperature <br> Range ${ }^{(1)}$ | Package |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

## Pin Configuration



6 -Pin $1.6 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ TMLF (MT) (Top View) MIC2095-1YMT/MIC2098-1YMT

$6-P i n 1.6 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ TMLF (MT) (Top View) MIC2097-1YMT / MIC2099-1YMT

$6-$ Pin $1.6 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ TMLF (MT) (Top View) MIC2095-2YMT/MIC2098-2YMT

$6-\mathrm{Pin} 1.6 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ TMLF (MT) (Top View) MIC2097-2YMT / MIC2099-2YMT

## Pin Description

| Pin Number | Pin Name | Pin Function |
| :---: | :---: | :--- |
| 1 | VOUT | Switch output (Output): The load being driven by the switch is connected to this pin. |
| 2 | NC | No Connect; Pin not used. |
| (MIC2095/MIC2098) | ILIMIT | Current Limit (Input): A resistor from this pin to ground sets the current limit value. See <br> the "setting ILMIIT" section for details on setting the resistor value. |
| 2 | FAULT/ | Fault status (Output): A logic low on this pin indicates the switch is in current limiting, or <br> has been shut down by the thermal protection circuit. This is an open-drain output <br> allowing logical OR'ing of FAULT/ outputs from multiple devices. |
| 3 | ENABLE | Switch Enable (Input): Logic high on this pin enables the switch. |
| 4 <br> (MIC2095-1/MIC2097-1/ <br> MIC2098-1/MIC2099-1) | ENABLE/ | Switch Enable (Input): Logic low on this pin enables the switch. |
| 4 <br> (MIC2095-2/MIC2097-2/ <br> MIC2098-2/MIC2099-2) | GNA | GND |
| 5 | VIN | Ground. <br> 6 <br> Power input (Input): This pin provides power to both the output power switch and the <br> internal control circuitry. |
| EP | Used to remove heat from die. Connect to ground. Use multiple vias to the ground <br> plane to minimize thermal impedance. See Applications Section for additional <br> information. |  |

Absolute Maximum Ratings ${ }^{(1)}$
Supply Voltage ( $\mathrm{V}_{\mathrm{IN}}$ ). $\qquad$ -0.3 V to 6.0 V
Output Voltage ( $\mathrm{V}_{\text {OUT }}$ ) ....................................... -0.3 V to $\mathrm{V}_{\text {IN }}$
FAULT Pin Voltage ( $\mathrm{V}_{\text {FAULT }}$ ) .............................. -0.3 V to $\mathrm{V}_{\mathrm{IN}}$
ENABLE Pin Voltage ( $\mathrm{V}_{\text {ENABLE }}$ ).......................... -0.3 V to $\mathrm{V}_{\text {IN }}$
ILIMIT Pin Voltage ( $\mathrm{V}_{\text {ILIMIT }}$ ) ................................ -0.3 V to $\mathrm{V}_{\text {IN }}$
Power Dissipation $\left(P_{D}\right) \ldots \ldots . . . . . . . . . . . . . . . . . . . . . .$. Internally Limited
Maximum Junction Temperature ( $\mathrm{T}_{\mathrm{J}}$ )......................... $150^{\circ} \mathrm{C}$
Storage Temperature (Ts)........................ $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Lead Temperature (soldering, 10sec.)....................... $260^{\circ} \mathrm{C}$
ESD HBM Rating (VOUT, GND) ${ }^{(3)}$................................. 4 kV


## Operating Ratings ${ }^{(2)}$

Supply Voltage ( $\mathrm{V}_{\mathrm{IN}}$ )........................................ 2.5 V to 5.5 V
ENABLE Pin Voltage ( $\mathrm{V}_{\text {ENAbLE }}$ ) .............................. 0 V to $\mathrm{V}_{\mathrm{IN}}$
FAULT Pin Voltage ( $\mathrm{V}_{\text {FAULT }}$ )................................... 0 V to $\mathrm{V}_{\mathrm{IN}}$
Ambient Temperature Range $\left(\mathrm{T}_{\mathrm{A}}\right) \ldots \ldots \ldots \ldots . .-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
Package Thermal Resistance ${ }^{(6)}$
$1.6 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ TMLF $\left(\theta_{\mathrm{JA}}\right) \ldots \ldots . . . . . . . . . . . . . . . . . . . . . . . .9^{\circ} \mathrm{C} / \mathrm{W}$

## Electrical Characteristics ${ }^{(4)}$

$\mathrm{V}_{\mathrm{IN}}=5 \mathrm{~V} ; \mathrm{C}_{\mathrm{IN}}=1 \mu \mathrm{~F} \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ unless noted, bold values indicate $-40^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq+85^{\circ} \mathrm{C}$.

| Symbol | Parameter | Condition | Min. | Typ. | Max. | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power Input Supply |  |  |  |  |  |  |
| $\mathrm{V}_{\text {IN }}$ | Input Voltage Range |  | 2.5 |  | 5.5 | V |
| $\mathrm{l}_{\mathrm{IN}}$ | Quiescent Supply Current ${ }^{(5)}$ | Switch = ON <br> Active Low Enable, $\mathrm{V}_{\mathrm{EN}}=0 \mathrm{~V}$ <br> Active High Enable, $\mathrm{V}_{\mathrm{EN}}=1.5 \mathrm{~V}$ |  | 80 | 300 | $\mu \mathrm{A}$ |
|  | Shutdown Current | $\begin{aligned} & \text { Switch = OFF } \\ & \text { Active Low Enable, } \mathrm{V}_{\mathrm{EN}}=1.5 \mathrm{~V} \end{aligned}$ |  | 8 | 15 | $\mu \mathrm{A}$ |
|  |  | Switch = OFF <br> Active High Enable, $\mathrm{V}_{\mathrm{EN}}=0.5 \mathrm{~V}$ |  | 0.1 | 5 | $\mu \mathrm{A}$ |
| UVLO ${ }_{\text {threshold }}$ | VIN UVLO Threshold | $V_{\text {IN }}$ Rising | 2 | 2.25 | 2.5 | V |
|  |  | $V_{\text {IN }}$ Falling | 1.9 | 2.15 | 2.4 | V |
|  | $V_{\text {IN }}$ UVLO Hysteresis |  |  | 100 |  | mV |
| Enable Control |  |  |  |  |  |  |
| $\mathrm{V}_{\text {EN }}$ | ENABLE Logic Level Low ${ }^{(5)}$ | $\mathrm{V}_{\text {IL(MAX }}$ |  |  | 0.5 | V |
|  | ENABLE Logic Level High ${ }^{(5)}$ | $\mathrm{V}_{\mathrm{IH} \text { (MIN) }}$ | 1.5 |  |  | V |
| $\mathrm{I}_{\mathrm{EN}}$ | ENABLE Bias Current | $0 \mathrm{~V} \leq \mathrm{V}_{\text {EN }} \leq 5 \mathrm{~V}$ |  | 0.1 | 5 | $\mu \mathrm{A}$ |
| ton_dLy | Output Turn-on Delay | $\begin{aligned} & R_{\mathrm{L}}=43 \Omega, C_{\mathrm{L}}=120 \mu \mathrm{~F} \\ & \mathrm{~V}_{\mathrm{EN}}=50 \% \text { to } \mathrm{V}_{\text {OUT }}=10 \% \end{aligned}$ |  | 1000 | 1500 | $\mu \mathrm{s}$ |
| toff_dLY | Output Turn-off Delay | $\begin{aligned} & \mathrm{R}_{\mathrm{L}}=43 \Omega, \mathrm{C}_{\mathrm{L}}=120 \mu \mathrm{~F} \\ & \mathrm{~V}_{\mathrm{EN}}=50 \% \text { to } \mathrm{V}_{\text {OUT }}=90 \% \end{aligned}$ |  |  | 700 | $\mu \mathrm{s}$ |
| $t_{\text {RISE }}$ | Output Turn-on rise time | $\begin{aligned} & R_{\mathrm{L}}=100 \Omega, \mathrm{C}_{\mathrm{LOAD}}=1 \mu \mathrm{~F} \\ & \mathrm{~V}_{\text {OUT }}=10 \% \text { to } 90 \% \end{aligned}$ | 500 | 1000 | 1500 | $\mu \mathrm{s}$ |
| Thermal Protection |  |  |  |  |  |  |
| OT Threshold | Over-temperature Shutdown | TJ Rising |  | 145 |  | ${ }^{\circ} \mathrm{C}$ |
|  |  | TJ Falling |  | 135 |  | ${ }^{\circ} \mathrm{C}$ |

## Electrical Characteristics (Continued)

$\mathrm{V}_{\mathrm{IN}}=5 \mathrm{~V} ; \mathrm{C}_{\mathrm{IN}}=1 \mu \mathrm{~F} \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ unless noted, bold values indicate $-40^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq+85^{\circ} \mathrm{C}$.

| Symbol | Parameter | Condition | Min. | Typ. | Max. | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Internal Switch |  |  |  |  |  |  |
| $\mathrm{R}_{\mathrm{DS}(\mathrm{ON})}$ | On Resistance | $\mathrm{V}_{\text {IN }}=5 \mathrm{~V}, \mathrm{I}_{\text {OUT }}=100 \mathrm{~mA}$ |  | 170 | 220 | $\mathrm{m} \Omega$ |
|  |  |  |  |  | 275 | $\mathrm{m} \Omega$ |
| ILEAK | Output Leakage Current | Switch $=$ OFF, $\mathrm{V}_{\text {OUt }}=0 \mathrm{~V}$ <br> Active Low Enable, $\mathrm{V}_{\mathrm{EN}}=1.5 \mathrm{~V}$ <br> Active High Enable, $\mathrm{V}_{\mathrm{EN}}=0 \mathrm{~V}$ |  | 0.1 | 10 | $\mu \mathrm{A}$ |
| Output Current Limit (MIC2095) |  |  |  |  |  |  |
| ILIMIT | Fixed Current Limit | $\mathrm{V}_{\text {OUT }}=0.8 \times \mathrm{V}_{\text {IN }}$ | 0.5 | 0.7 | 0.9 | A |
| Output Current Limit (MIC2098) |  |  |  |  |  |  |
| ILIMIT | Fixed Current Limit | $\mathrm{V}_{\text {OUT }}=0.8 \times \mathrm{V}_{\text {IN }}$ | 0.9 | 1.1 | 1.5 | A |
| Output Current Limit (MIC2097, MIC2099) |  |  |  |  |  |  |
| CLF | Variable Current Limit Factors | $\mathrm{I}_{\text {OUT }}=1.1 \mathrm{~A}, \mathrm{~V}_{\text {OUT }}=0.8 \times \mathrm{V}_{\text {IN }} ; \mathrm{V}_{\text {IN }}=2.5 \mathrm{~V}$ | 175 | 215 | 263 | V |
|  |  | lout $=0.5 \mathrm{~A}, \mathrm{~V}_{\text {OUT }}=0.8 \times \mathrm{V}_{\text {IN }} ; \mathrm{V}_{\text {IN }}=2.5 \mathrm{~V}$ | 152 | 206 | 263 | V |
|  |  | $\mathrm{l}_{\text {OUT }}=0.2 \mathrm{~A}, \mathrm{~V}_{\text {OUT }}=0.8 \times \mathrm{V}_{\text {IN }} ; \mathrm{V}_{\text {IN }}=2.5 \mathrm{~V}$ | 138 | 200 | 263 | V |
|  |  | $\mathrm{I}_{\text {OUT }}=0.1 \mathrm{~A}, \mathrm{~V}_{\text {OUT }}=0.8 \times \mathrm{V}_{\text {IN }} ; \mathrm{V}_{\text {IN }}=2.5 \mathrm{~V}$ | 121 | 192 | 263 | V |
| Kickstart ${ }^{\text {TM }}$ Current Limit (MIC2097) |  |  |  |  |  |  |
| ILIMIT_2nd | Secondary Current Limit | $\mathrm{V}_{\text {IN }}=2.5 \mathrm{~V} ; \mathrm{V}_{\text {OUT }}=0 \mathrm{~V}$ |  | 1.5 |  | A |
| $\mathrm{t}_{\text {_LIMIT }}$ | Duration of Kickstart ${ }^{\text {TM }}$ Current Limit | $\mathrm{V}_{\text {IN }}=2.5 \mathrm{~V}$ | 77 | 105 | 192 | ms |
| Fault Flag |  |  |  |  |  |  |
| $V_{\text {FAuLt }}$ | Fault Flag Output Voltage | $\mathrm{loL}=10 \mathrm{~mA}$ |  | 0.25 | 0.4 | V |
|  | Fault Flag Off Current | $\mathrm{V}_{\text {FAULT/ }}=5 \mathrm{~V}$ |  | 0.01 | 1 | $\mu \mathrm{A}$ |
| Fault Delay (MIC2095, MIC2098, MIC2099) |  |  |  |  |  |  |
| $t_{\text {D_FAULT }}$ | Delay before asserting or releasing FAULT/ | Time from current limiting (VOUT $=0.4 \mathrm{x}$ $\mathrm{V}_{\mathrm{IN}}$ ) to FAULT/ state change | 20 | 32 | 49 | ms |
| Fault Delay (MIC2097) |  |  |  |  |  |  |
| $t_{\text {D_FAULT }}$ | Delay before asserting or releasing FAULT/ | Time from current limiting $\left(\mathrm{V}_{\text {OUT }}=0.8 \mathrm{x}\right.$ $\mathrm{V}_{\mathrm{IN}}$ ) to $\mathrm{FAULT} /$ state change; $\mathrm{V}_{\mathrm{IN}}=2.5 \mathrm{~V}$ | 77 | 105 | 192 | ms |

## Notes:

1. Exceeding the absolute maximum rating may damage the device.
2. The device is not guaranteed to function outside its operating rating.
3. Devices are ESD sensitive. Handling precautions recommended. Human body model, 1.5 k in series with 100 pF .
4. Specifications for packaged product only.
5. Check the Ordering Information section to determine which parts are Active High or Active Low.
6. Requires proper thermal mounting to achieve this performance.

## Timing Diagrams



Rise and Fall Times


Switching Delay Times

## Typical Characteristics














## Typical Characteristics (Continued)














## Typical Characteristics (Continued)






## Functional Characteristics



Time ( $2.0 \mathrm{~ms} / \mathrm{div}$ )


Time ( $100 \mathrm{~ms} / \mathrm{div}$ )



## Functional Characteristics (Continued)





## Functional Characteristics (Continued)



Time (100ms/div)



2099 Output Recovery from Short Circuit and Fault/ Response


## Functional Characteristics (Continued)



## Functional Characteristics (Continued)



## Functional Diagram



MIC2095/97/98/99 Functional Diagram

## Functional Description and Application Information

## $\mathrm{V}_{\text {IN }}$ and $\mathrm{V}_{\text {out }}$

$\mathrm{V}_{\text {IN }}$ is both the power supply connection for the internal circuitry driving the switch and the input (Source connection) of the power MOSFET switch. $\mathrm{V}_{\text {Out }}$ is the Drain connection of the power MOSFET and supplies power to the load. In a typical circuit, current flows from $V_{\text {IN }}$ to $V_{\text {OUt }}$ toward the load. Since the switch is bidirectional when enabled, if $\mathrm{V}_{\text {OUt }}$ is greater than $\mathrm{V}_{\mathrm{IN}}$, current will flow from $\mathrm{V}_{\text {Out }}$ to $\mathrm{V}_{\text {IN }}$.
When the switch is disabled, current will not flow to the load, except for a small unavoidable leakage current of a few micro amps. However, should $V_{\text {OUt }}$ exceed $\mathrm{V}_{\text {IN }}$ by more than a diode drop $(\sim 0.6 \mathrm{~V})$, while the switch is disabled, current will flow from output to input via the power MOSFET's body diode. When the switch is enabled, current can flow both ways, from $\mathrm{V}_{\text {IN }}$ to $\mathrm{V}_{\text {OUT }}$, or $V_{\text {Out }}$ to $\mathrm{V}_{\text {IN }}$.

## $\mathrm{C}_{\mathrm{IN}}$

A minimum $1 \mu \mathrm{~F}$ bypass capacitor positioned as close as possible to the $\mathrm{V}_{\mathbb{I}}$ and GND pins of the switch is both good design practice and required for proper operation of the switch. This will control supply transients and ringing. Without a sufficient bypass capacitor, large current surges or a short may cause sufficient ringing on $\mathrm{V}_{\mathrm{IN}}$ (from supply lead inductance) to cause erratic operation of the switch's control circuitry. For best performance a good quality, low-ESR ceramic capacitor is recommended.
An additional $22 \mu \mathrm{~F}$ (or greater) capacitor, positioned close to the $\mathrm{V}_{\mathrm{IN}}$ and GND pins of the switch is necessary if the distance between a larger bulk capacitor and the switch is greater than 3 inches. This additional capacitor limits input voltage transients at the switch caused by fast changing input currents that occur during a fault condition, such as current limit and thermal shutdown.
When bypassing with capacitors of $10 \mu \mathrm{~F}$ and up, it is good practice to place a smaller value capacitor in parallel with the larger to handle the high frequency components of any line transients. Values in the range of $0.1 \mu \mathrm{~F}$ to $1 \mu \mathrm{~F}$ are recommended. Again, good quality, low-ESR capacitors, preferably ceramic, should be chosen.

## $\mathrm{C}_{\text {out }}$

An output capacitor is recommended to reduce ringing and voltage sag on the output during a transient condition. A value between $1 \mu \mathrm{f}$ and $10 \mu \mathrm{f}$ is recommended, however, larger values can be used.

## Limitations on $\mathrm{C}_{\text {out }}$

The part may enter current limit when turning on with a large output capacitance. This is an acceptable condition, however, if the part remains in current limit for a time greater than $\mathrm{t}_{\mathrm{D} \text { _falt }}$, the FAULT pin will assert low. The maximum value of COUT may be approximated by the following equation:

$$
\begin{equation*}
\mathrm{C}_{\text {OUT_MAX }}=\frac{\text { LIMIT_MIN } \times \mathrm{t}_{\text {D_FAULT_MIN }}}{\mathrm{V}_{\text {IN_MAX }^{\prime}}} \tag{Eq. 1}
\end{equation*}
$$

Where: $I_{\text {limit_min }}$ and $t_{\text {d_faultamin }}$ are the minimum specified values listed in the Electrical Characteristic table and $\mathrm{V}_{\mathbb{I N}^{\prime} \max }$ is the maximum input voltage to the switch.

## Current Sensing and Limiting

The current limiting switches protect the system power supply and load from damage by continuously monitoring current through the on-chip power MOSFET. Load current is monitored by means of a current mirror in parallel with the power MOSFET switch. Current limiting is invoked when the load exceeds the overcurrent threshold. When current limiting is activated the output current is constrained to the limit value, and remains at this level until either the load/fault is removed, the load's current requirement drops below the limiting value, or the switch goes into thermal shutdown.

## Kickstart ${ }^{\text {TM }}$

The MIC2097 has a Kickstart feature that allows higher momentary current surges before the onset of current limiting. This permits dynamic loads, such as small disk drives or portable printers to draw the inrush current needed to overcome inertial loads without sacrificing system safety. The Kickstart parts differ from the nonKickstart parts which more rapidly limit load current, potentially starving a motor and causing the appliance to stall or stutter.
During the Kickstart delay period, (typically 105ms), a secondary current limit (nominally set at 1.5 A ), is in effect. If the load demands a current in excess the secondary limit, Kickstart parts act immediately to restrict output current to the secondary limit for the duration of the Kickstart period. After this time the Kickstart parts revert to their normal current limit. An example of Kickstart operation is in Figure 1.
Kickstart may be over-ridden by the thermal protection circuit and if sufficient internal heating occurs, Kickstart will be terminated and the output switch will be turned off. After the parts cools, if the load is still present $\mathrm{I}_{\text {OUt }} \rightarrow$ limit, not limit_2nd.


Figure 1. MIC2097 Kickstart Operation
Figure 1 Label Key:
A. The MIC2097 is enabled into an excessive load (slew-rate limiting not visible at this time scale) The initial current surge is limited by either the overall circuit resistance and power-supply compliance, or the secondary current limit, whichever is less.
B. Ron of the power FET increases due to internal heating.
C. Kickstart period.
D. Current limiting initiated. FAULT/ goes low.
E. $V_{\text {OUt }}$ is non-zero (load is heavy, but not a dead short where $\mathrm{V}_{\text {OUt }}=0 \mathrm{~V}$. Limiting response will be the same for dead shorts).
F. Thermal shutdown followed by thermal cycling.
G. Excessive load released, normal load remains. MIC2097 drops out of current limiting.
H. FAULT/ delay period followed by FAULT/ going HIGH.

## Enable Input

The ENABLE pin is a logic level compatible input which turns on or off the main MOSFET switch. There are two versions of each device. The -1 version has an active high (ENABLE) and the -2 version has an active low (ENABLE/).

## Fault Output

The FAULT/ is an N-channel open-drain output, which is asserted (LOW true) when the device either begins current limiting or enters thermal shutdown. The FAULT/ signal asserts after a brief delay period in order to filter out very brief over current conditions. After an overcurrent or over-temperature fault clears, the FAULT/ pin remains asserted (low) for the delay period.
The FAULT/output is open-drain and must be pulled HIGH with an external resistor. The FAULT/ signal may be wire-OR'd with other similar outputs, sharing a single
pull-up resistor. FAULT/ may be tied to a pull-up voltage source which is less than or equal to $\mathrm{V}_{\operatorname{IN}}$.

## Soft-Start Control

Large capacitive loads can create significant inrush current surges when charged through the current limiting switch. When the switch is enabled, the built-in soft-start limits the initial inrush current by slowly turning on the output.

## Power Dissipation and Thermal Shutdown

Thermal shutdown is used to protect the current limiting switch from damage should the die temperature exceed a safe operating temperature. Thermal shutdown shuts off the output MOSFET and asserts the FAULT/ output if the die temperature reaches $145^{\circ} \mathrm{C}$ (typical).
The switch will automatically resume operation when the die temperature cools down to $135^{\circ} \mathrm{C}$. If resumed operation results in reheating of the die, another shutdown cycle will occur and the switch will continue cycling between ON and OFF states until the reason for the overcurrent condition has been resolved.
Depending on PCB layout, package type, ambient temperature, etc., hundreds of milliseconds may elapse from the time a fault occurs to the time the output MOSFET will be shut off. This delay is caused because of the time it takes for the die to heat after the fault condition occurs.
Power dissipation depends on several factors such as the load, PCB layout, ambient temperature, and supply voltage. Calculation of power dissipation can be accomplished by the following equation:

$$
\begin{equation*}
\mathrm{P}_{\mathrm{D}}=\mathrm{R}_{\mathrm{DS}(\mathrm{ON})} \times\left(\mathrm{l}_{\mathrm{OUT}}\right)^{2} \tag{Eq. 2}
\end{equation*}
$$

To relate this to junction temperature, the following equation can be used:

$$
\begin{equation*}
T_{J}=P_{D} \times R_{\theta(J-A)}+T_{A} \tag{Eq. 3}
\end{equation*}
$$

Where $\mathrm{T}_{\mathrm{J}}=$ junction temperature, $\mathrm{T}_{\mathrm{A}}=$ ambient temperature, and $\mathrm{R}_{\theta(J-A)}$ is the thermal resistance of the package.
In normal operation, excessive switch heating is most often caused by an output short circuit. If the output is shorted, when the switch is enabled, the switch limits the output current to the maximum value. The heat generated by the power dissipation of the switch continuously limiting the current may exceed the package and PCB's ability to cool the device and the switch will shut down and signal a fault condition. Please see the Fault Output description in the previous page for more details on the FAULT/ output. After the switch
shuts down, and cools, it will re-start itself if the Enable signal retains true (high on the ENABLE parts, low on the ENABLE/ parts).
In Figure 2, die temperature is plotted against lout assuming a constant ambient temperature of $85^{\circ} \mathrm{C}$. The plot also assumes the maximum specified switch resistance at high temperature.


Figure 2. Die Temperature vs. Iout

## Setting $\mathrm{I}_{\text {Limit }}$

The current limit of the MIC2097 and MIC2099 parts are user programmable and controlled by a resistor connected between the $I_{\text {LIMIT }}$ pin and Ground. The value of the current limit resistor is determined by the following equations:

$$
\begin{equation*}
\mathrm{I}_{\mathrm{LIMIT}}=\frac{\text { CurrentLimitFactor(CLF) }}{\mathrm{R}_{\text {LIMIT }}} \tag{Eq. 4}
\end{equation*}
$$

or

$$
\begin{equation*}
\mathrm{R}_{\mathrm{LIMIT}}=\frac{\text { CurrentLimitFactor(CLF) }}{\mathrm{I}_{\mathrm{LIMIT}}} \tag{Eq. 5}
\end{equation*}
$$

The Current-Limit Factor (CLF) is a number that is characteristic to the MIC2097/9 switches. The CLF is a product of the current-setting resistor value, and the desired current-limit value. Please note that the CLF varies with the current output current, so caution is necessary to use the correct CLF value for the current that you intend to use the part at. For example: If one wishes to set a $l_{\text {LIMIT }}=1.1 \mathrm{~A}$, looking in the electrical specifications we will find CLF at $\mathrm{I}_{\text {LIMIT }}=1.1 \mathrm{~A}$, as noted in Table 1.

| Min. | Typ. | Max. | Units |
| :---: | :---: | :---: | :---: |
| 175 | 215 | 263 | V |

Table 1. $C L F$ at $I_{\text {LIMIT }}=1.1 \mathrm{~A}$

For the sake of this example, the typical value of CLF at an $\mathrm{I}_{\text {OUt }}$ of 1.1 A is 215 V . Applying Equation 5 :

$$
\begin{equation*}
\mathrm{R}_{\mathrm{LIMIT}}(\Omega)=\frac{215 \mathrm{~V}}{1.1 \mathrm{~A}}=195 \Omega \tag{Eq. 6}
\end{equation*}
$$

Choose $\mathrm{R}_{\text {LIMIT }}=196 \Omega$ (the closest standard $1 \%$ value) Designers should be aware that variations in the measured $\mathrm{I}_{\text {LIMIT }}$ for a given $\mathrm{R}_{\text {LIMIT }}$ resistor, will occur because of small differences between individual ICs (inherent in silicon processing) resulting in a spread of $l_{\text {Limit }}$ values. In the example above we used the typical value of CLF to calculate $\mathrm{R}_{\text {LIMIT }}$. We can determine $I_{\text {LImit's }}$ spread by using the minimum and maximum values of CLF and the calculated value of $\mathrm{R}_{\text {LIмIт }}$ :

$$
\begin{equation*}
\mathrm{I}_{\text {LIMIT_MIN }}=\frac{175 \mathrm{~V}}{196 \Omega}=0.89 \mathrm{~A} \tag{Eq. 7}
\end{equation*}
$$

$$
\begin{equation*}
\mathrm{I}_{\text {LIMIT_MAX }}=\frac{263 \mathrm{~V}}{196 \Omega}=1.34 \mathrm{~A} \tag{Eq. 8}
\end{equation*}
$$

Giving us a maximum $\mathrm{I}_{\text {LIMIT }}$ variation of:

| ILIMIT_MIN $^{\text {LIMIT_TYP }}$ | limit_max $^{\text {LIMA }}$ |  |
| :--- | :--- | :--- |
| $0.89 \mathrm{~A}(-19 \%)$ | 1.1 A | $1.34 \mathrm{~A}(+22 \%)$ |

For convenience, Table 2 lists the resistance values for the $R_{\text {SET }}$ pin, for various current limit values.

| Nominal <br> $I_{\text {LImit }}$ | $\mathrm{R}_{\text {LIIIIt }}$ | lumit_min | Ilimit_max |
| :---: | :---: | :---: | :---: |
| 0.1 A | 1920 | 0.063 | 0.137 |
| 0.2A | 1000 | 0.138 | 0.263 |
| 0.3A | 672 | 0.211 | 0.391 |
| 0.4 A | 508 | 0.288 | 0.517 |
| 0.5A | 412 | 0.369 | 0.638 |
| 0.6A | 344 | 0.448 | 0.764 |
| 0.7 A | 298 | 0.533 | 0.884 |
| 0.8A | 263 | 0.620 | 1.002 |
| 0.9A | 235 | 0.709 | 1.118 |
| 1.0A | 213 | 0.801 | 1.233 |
| 1.1A | 195 | 0.895 | 1.346 |

Table 2. MIC2097 and MIC2099 Rlimit Table

## $I_{\text {LIMIT }}$ vs. I Iout Measured

When in current limit, the switches are designed to act as a constant-current source to the load. As the load tries to pull more than the maximum current, $\mathrm{V}_{\text {OUT }}$ drops and the input-to-output voltage differential increases. As the ( $\mathrm{V}_{\text {IN }}-\mathrm{V}_{\text {OUT }}$ ) voltage differential increases, the IC internal temperature also increases. To limit the IC's power dissipation, the current limit is reduced as a function of output voltage.
This folding back of $\mathrm{I}_{\text {LIMIT }}$ can be generalized by plotting $\mathrm{I}_{\text {Limit }}$ as a function of $\mathrm{V}_{\text {OUt }}$, as shown in Figures 3 and 4. The slope of $\mathrm{V}_{\text {OUt }}$ between $\mathrm{l}_{\text {out }}=0 \mathrm{~V}$ and $\mathrm{l}_{\text {out }}=\mathrm{I}_{\text {Limit }}$ (where $\mathrm{I}_{\text {LIMIT }}$ is a normalized 1 A ) is determined by $\mathrm{R}_{\mathrm{ON}}$ of the switch and $\mathrm{I}_{\text {LIMIT }}$.


Figure 3. Normalized Output Current vs. Output Voltage


Figure 4. Normalized Output Current vs. Output Voltage

When measuring $l_{\text {out }}$ it is important to remember voltage dependence, otherwise the measurement data may appear to indicate a problem when none really exists. This voltage dependence is illustrated in Figures 5 and 6.

In Figure 5, output current is measured as $\mathrm{V}_{\text {OUt }}$ is pulled below $\mathrm{V}_{\mathbb{N}}$, with the test terminating when $\mathrm{V}_{\text {Out }}$ is 1 V below $\mathrm{V}_{\text {IN }}$. Observe that once $\mathrm{I}_{\text {LIMIT }}$ is reached $\mathrm{l}_{\text {out }}$ remains constant throughout the remainder of the test. In Figure 6 this test is repeated but with $\left(\mathrm{V}_{\mathbb{I N}}-\mathrm{V}_{\mathrm{OUT}}\right)$ is 4 V .


Figure 5. lout in Current Limiting for $\mathrm{V}_{\text {out }}=4 \mathrm{~V}$


Figure 6. lout $^{\text {in }}$ Current Limiting for $\mathrm{V}_{\text {out }}=1 \mathrm{~V}$

## Under Voltage Lock Out (UVLO)

The switches have an Under Voltage Lock Out (UVLO) feature that will shut down the switch in a reproducible manner when the input power supply voltage goes too low. The UVLO circuit disables the output until the supply voltage exceeds the UVLO threshold. Hysteresis in the UVLO circuit prevents noise and finite circuit impedance from causing chatter during turn-on and turnoff. While disable by the UVLO circuit, the output switch (power MOSFET) is OFF and no circuit functions, such as FAULT/ or ENABLE, are considered to be valid or operative.

## Typical Application Schematics



Figure 7. MIC2095-1 or MIC2098-1 Typical Schematic
Note: MIC2095-1 and MIC2098-1; R5=NF; EN pin uses R4 (pull-up resistor to $\mathrm{V}_{\mathbb{I}}$ ) to enable the output without an external enable signal. MIC2095-2 and MIC2098-2; R4=NF; EN/ pin uses R5 (pull-down resistor to GND) to enable the output without an external enable signal.


Figure 8. MIC2097-1 Typical S̄chematic
Note: MIC2097-1; R5=NF; EN pin uses R4 (pull-up resistor to $V_{\text {IN }}$ ) to enable the output without an external enable signal. MIC2097-2; R4=NF; EN/ pin uses R5 (pull-down resistor to GND) to enable the output without an external enable signal.


Figure 9. MIC2099-1 Schematic
Note: MIC2099-1; R5=NF; EN pin uses R4 (pull-up resistor to $V_{I N}$ ) to enable the output without an external enable signal. MIC2099-2; R4=NF; EN/ pin uses R5 (pull-down resistor to GND) to enable the output without an external enable signal.

## Evaluation Board Schematic



Figure 10. Schematic of MIC209X Evaluation Board

## Notes:

1. Evaluation board is used for all parts.
2. Part numbering scheme is $209 \mathrm{X}-\mathrm{Y}$ where X is the place holder for the last number (i.e. MIC2095, MIC2097, MIC2098 or MIC2099) and Y is the polarity of the enable signal ( -1 indicates active high logic and -2 indicates active low logic).
3. MIC209X-1 EN pin only requires $R 4$ (pull-up resistor to $V_{I N}$ ) to enable the output without an external enable signal.
4. MIC209X-2 EN/ pin only requires R3 (pull-down resistor-to-GND) to enable the output without an external enable signal.
5. R1 is NF (no fill) with the MIC2095 (fixed current limit).

## MIC209x Bill of Materials

| Item | Part Number | Manufacturer | Description | Qty. |
| :---: | :---: | :---: | :---: | :---: |
| C1 | 08056D106MAT2A | $\mathrm{AV} \mathrm{X}^{(1)}$ | Ceramic Capacitor, 10 1 F, 6.3V, X5R | 1 |
| C2 | 06033D105MAT2A | $A V X^{(1)}$ | Ceramic Capacitor, $1 \mu \mathrm{~F}, 25 \mathrm{~V}$, X5R | 1 |
| C3 | 0805D226MAT2A | AVX ${ }^{(1)}$ | Ceramic Capacitor, 22 $\mu \mathrm{F}, 6.3 \mathrm{~V}$, X5R | 1 |
| C4 |  |  | 120 F ( (optional) | 0 |
| $\mathrm{R} 1^{(4)}$ | CRCW06032000FRT1 | Vishay Dale ${ }^{(2)}$ | Resistor, 200 (0603 size), 1\% | 1 |
| R2, R3, R4 | CRCW06031002FRT1 | Vishay Dale ${ }^{(2)}$ | Resistor, 10k (0603 size), 1\% | 3 |
| U1 | MIC2095-1YMT | Micrel, Inc. ${ }^{(3)}$ | Current-Limiting Power Distribution Switch - 0.5A Fixed Current Limit - Active High Enable | 1 |
| U1 | MIC2095-2YMT | Micrel, Inc. ${ }^{(3)}$ | Current-Limiting Power Distribution Switch - 0.5A Fixed Current Limit - Active Low Enable | 0 |
| U1 | MIC2097-1YMT | Micrel, Inc. ${ }^{(3)}$ | Current-Limiting Power Distribution Switch Adjustable Current Limit with Kickstart - Active High Enable | 0 |
| U1 | MIC2097-2YMT | Micrel, Inc. ${ }^{(3)}$ | Current-Limiting Power Distribution Switch Adjustable Current Limit with Kickstart - Active Low Enable | 0 |
| U1 | MIC2098-1YMT | Micrel, Inc. ${ }^{(3)}$ | Current-Limiting Power Distribution Switch - 0.9A Fixed Current Limit - Active High Enable | 0 |
| U1 | MIC2098-2YMT | Micrel, Inc. ${ }^{(3)}$ | Current-Limiting Power Distribution Switch - 0.9A Fixed Current Limit - Active Low Enable | 0 |
| U1 | MIC2099-1YMT | Micrel, Inc. ${ }^{(3)}$ | Current-Limiting Power Distribution Switch Adjustable Current Limit - Active High Enable | 0 |
| U1 | MIC2099-2YMT | Micrel, Inc. ${ }^{(3)}$ | Current-Limiting Power Distribution Switch Adjustable Current Limit - Active Low Enable | 0 |

Notes:

1. AVX: www.avx.com.
2. Vishay: www.vishay.com.
3. Micrel, Inc.: www.micrel.com.
4. May be omitted when used with the MIC2095 or MIC2098 (fixed current limit).

## PCB Layout Recommendations



Figure 11. MIC209X Evaluation Board Top Layer


Figure 12. MIC209X Evaluation Board Bottom Layer

## Package Information



NDTE

1. ALL DIMENSIDNS ARE IN MILLIMETERS.
2. MAX. PACKAGE WARPAGE IS 0.05 mm .
3. MAXIMUM ALLIWABE BURRS IS 0.076 mm IN ALL DIRECTIDNS.
. PIN \#1 ID DN TDP WILL BE LASER/INK MARKED.
DIMENSIDN APPLIES TD METALIZED TERMINAL AND IS MEASURED BETWEEN 0.20 AND 0.25 mm FRDM TERMINAL TIP.
4. APPLIED QNLY FDR TERMINALS.

A APPLIED FQR EXPISED PAD AND TERMINALS.

SIDE VIEW

6-Pin $1.6 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ TMLF (MT)

## Recommended Landing Pattern

## Recommended Land Pattern for MLF 1.6x1.6 6 Lead

LP \# MLF1616D-6LD-LP-1<br>All units are in mm<br>Tolerance $\pm 0.05$ if not noted



Red circle indicates Thermal Via. Size should be $300-350 \mathrm{~mm}$ in diameter and it should be connected to GND plane for maximum thermal performance.

6-Pin $1.6 \mathrm{~mm} \times 1.6 \mathrm{~mm}$ TMLF (MT)

## MICREL, INC. 2180 FORTUNE DRIVE SAN JOSE, CA 95131 USA <br> TEL +1 (408) 944-0800 FAX +1 (408) 474-1000 WEB http://www.micrel.com

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